

μ Mote: Enabling Passive Chirp De-spreading and μ W-level Long-Range Downlink for backscatter Devices

Yihang Song¹, Li Lu¹, Jiliang Wang², Chong Zhang¹, Hui Zheng¹,
Shen Yang¹, Jinsong Han³, and Jian Li¹

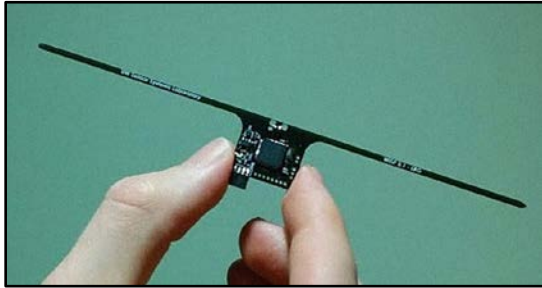
¹University of Electronic Science and Technology of China

²Tsinghua University

³Zhejiang University

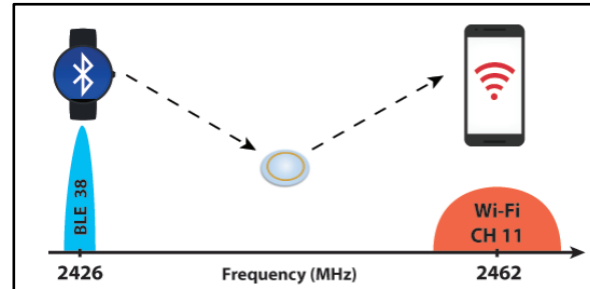


Conventional Backscatter Devices



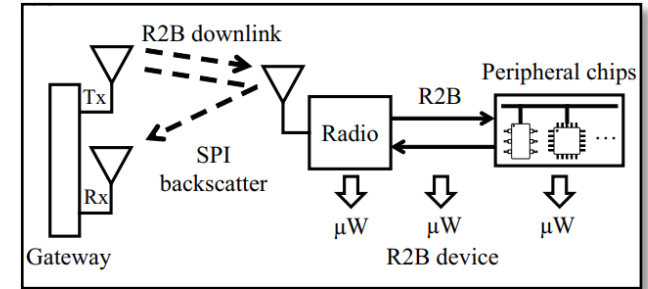
WISP Platform

IEEE TIM (2008)



Inter-Technology
Backscatter

SIGCOMM 2016



Internet-of-Microchips

MobiCom 2020



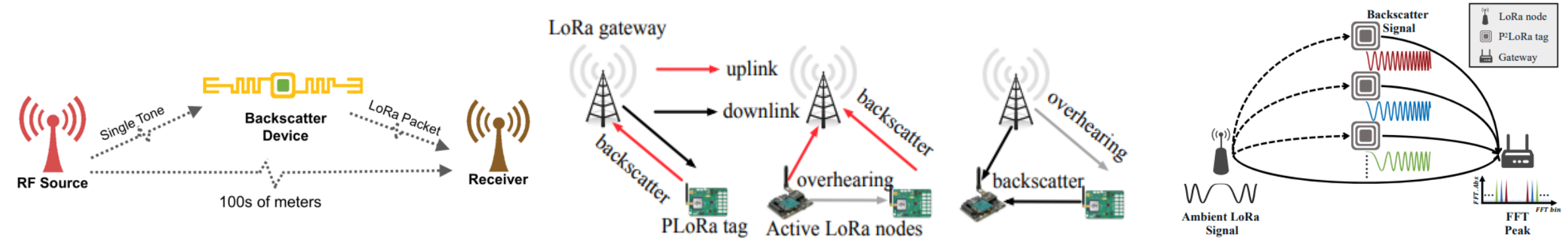
μ W-level low-power communication



Short range (<20 meters)

CSS-based Long-range Backscatter

CSS=Chirp Spread Spectrum



LoRa Backscatter

Power:	9.25μW
Uplink:	2.8km
Downlink:	5m

PLoRa

Power:	220μW
Uplink:	1.1km
Downlink:	<1m

P²IoRa

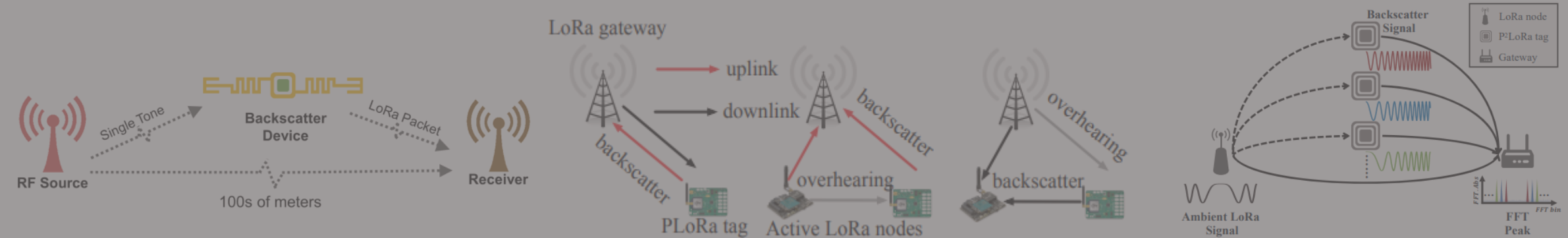
Power:	320μW
Uplink:	2.2km
Downlink:	<1m

😊 CSS significantly increases the uplink range with μW -level power

 The downlink range remains limited

CSS-based Long-range Backscatter

CSS=Chirp Spread Spectrum



Can we achieve **CSS on the downlink** to extend the receiving range of **low-power backscatter devices**?

Downlink: 5m

Downlink: <1m

Downlink: <1m

😊 CSS significantly increases the uplink range with μW -level power

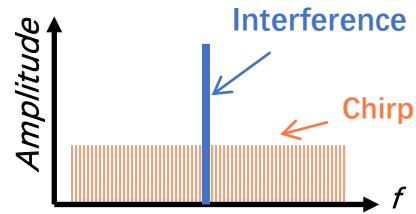


The downlink range remains limited

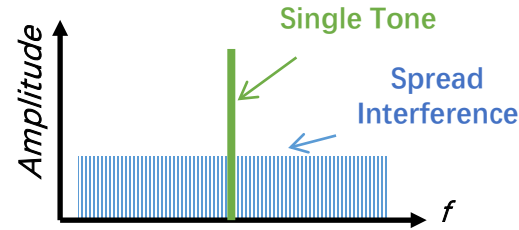
CSS communication principle

Basic idea of CSS

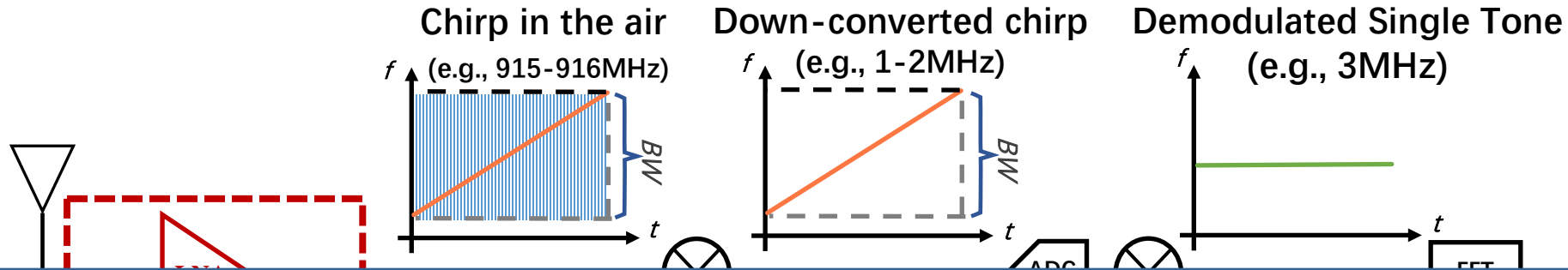
Before De-spreading



After De-spreading



De-spreading procedures



Chirp de-spreading and RF signal amplification consumes
unaffordable **Milliwatts of Power**

mWs

Two challenges

Challenge 1:

How to de-spread chirp to combat interference with extremely low power?

Challenge 2:

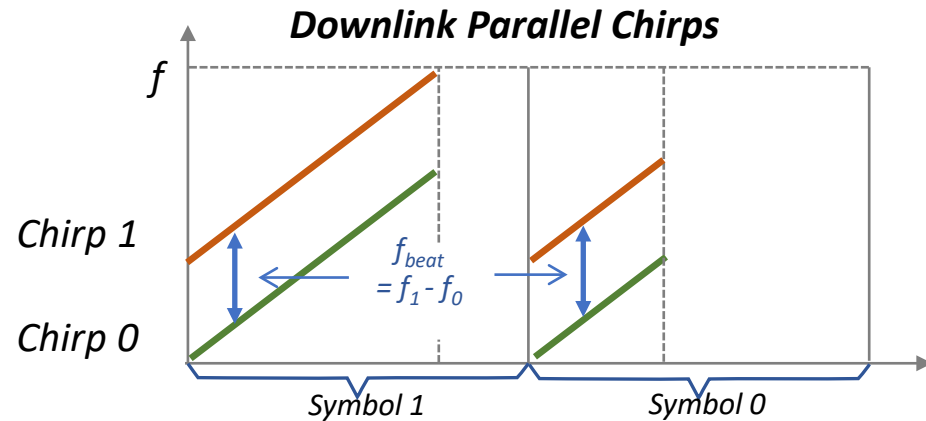
How to raise the signal amplitude with extremely low power?

Challenge 3

(will be discussed later)

Passive Chirp De-spreading Method

Basic idea: **upload** the power consuming carrier generation function

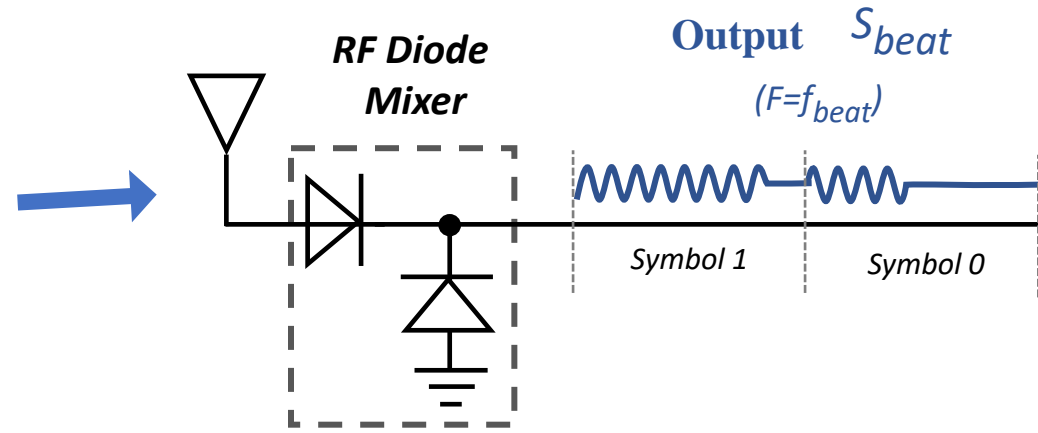


Chirp 0: the chirp signal to be de-spread

Chirp 1: plays the role of
carrier signal

+

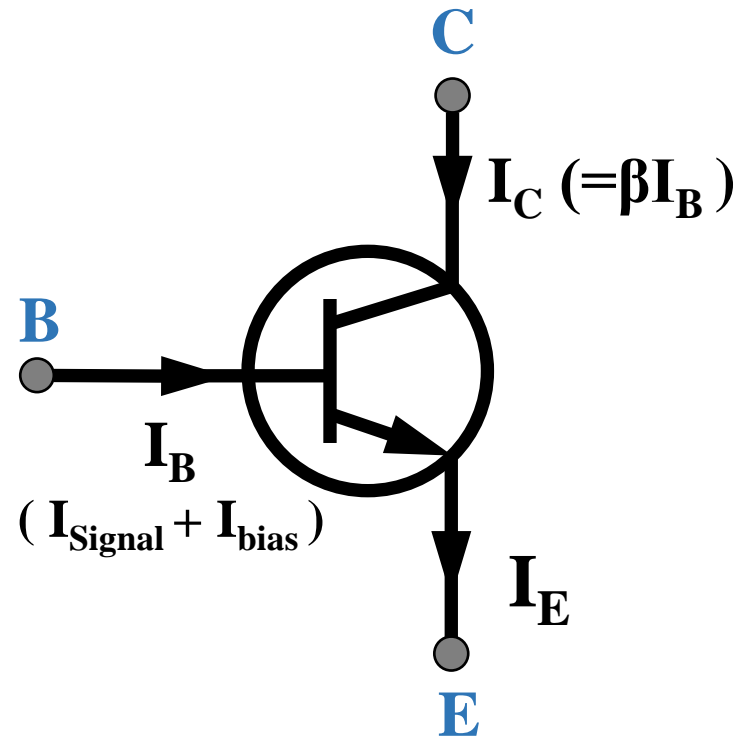
de-spreading signal (similar to the down-chirp
signal)



Using RF diode as mixer
for circuit simplicity

LNA principle analysis

Commercial
LNA IC



$$I_E = I_C + I_B$$

$$I_B = I_{Bias} + I_{Signal}$$

(100s of μA) (1-10 μA)

$$I_E \approx \beta I_B \quad (\beta \gg 1)$$
$$= \beta I_{Bias} + \beta I_{Signal}$$

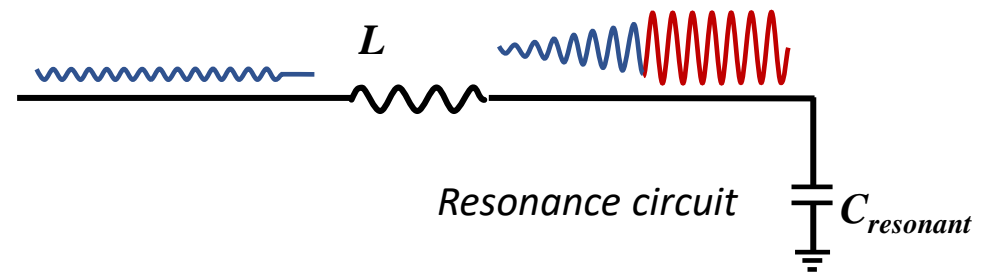
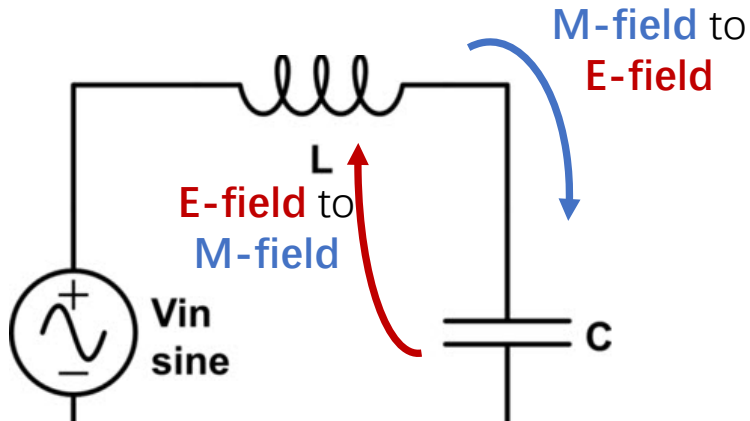
(10s of mA) (10s of μA)
(Unused) (with data)

Unused βI_{Bias} can waste 99% of power → Power Consumption: mW-level

Magnify Signals by Accumulating Energy

Our Solution: LC resonance circuit

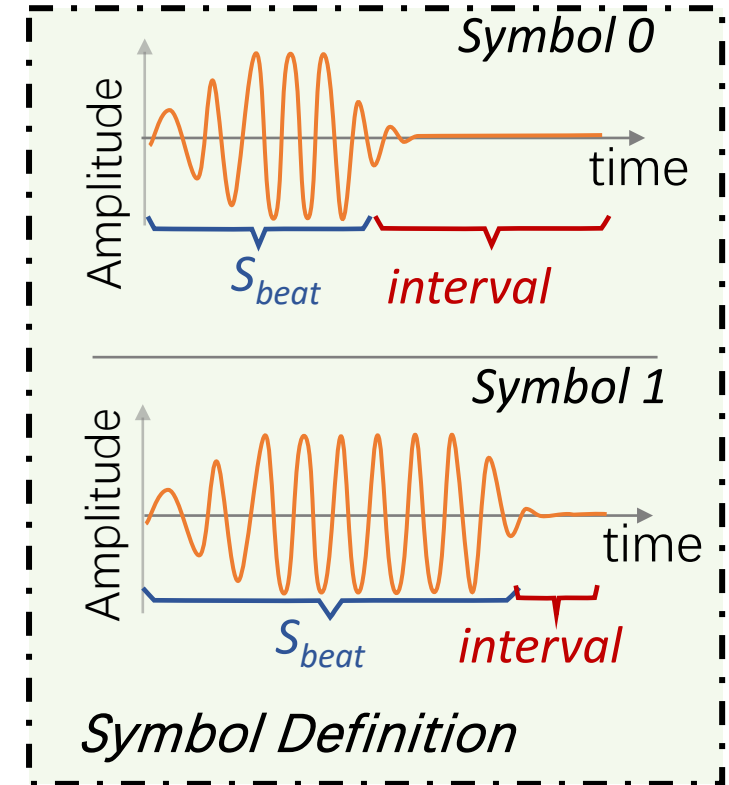
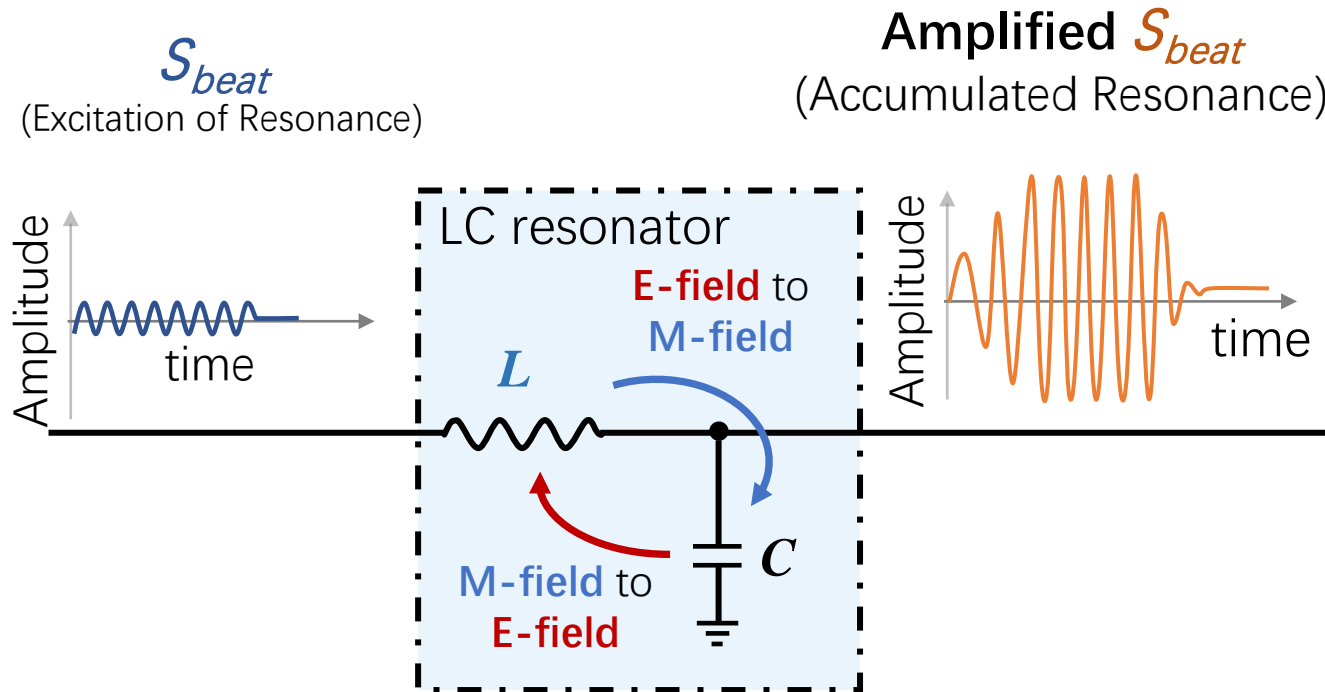
- ① Energy periodically transitions between **E** and **M** fields, creating **resonance**
- ② **Signal energy can be accumulated** in the form of resonance,
- ③ And Hence the signal **amplitude is magnified**



**Signal characteristics (e.g. frequency, phase, amplitude) are distorted.
How can the signal information be preserved during the accumulation?**

Magnify Signals by Accumulating Energy

Encoding data with the duration of resonance



Challenge 3

How to decode the symbol with low power?

Conventional solution: ADC-based sampling

A symbol to be decoded
(Blue dots represent sampling points)

- Sampling rate \rightarrow obey Nyquist's theorem
- Sampling operation numbers for each symbol : 80-120

Is it possible to decode a symbol with
a single integration operation?

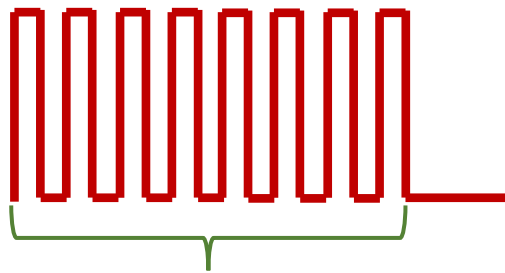
Sample Amplification and Integration

- Quantization (using inverse integration)
- Digital Output

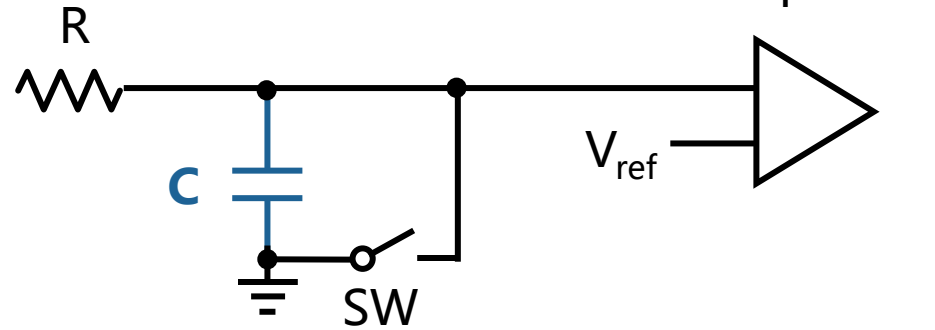
Typical power of ADC sampling: 100s of μW

Basic idea

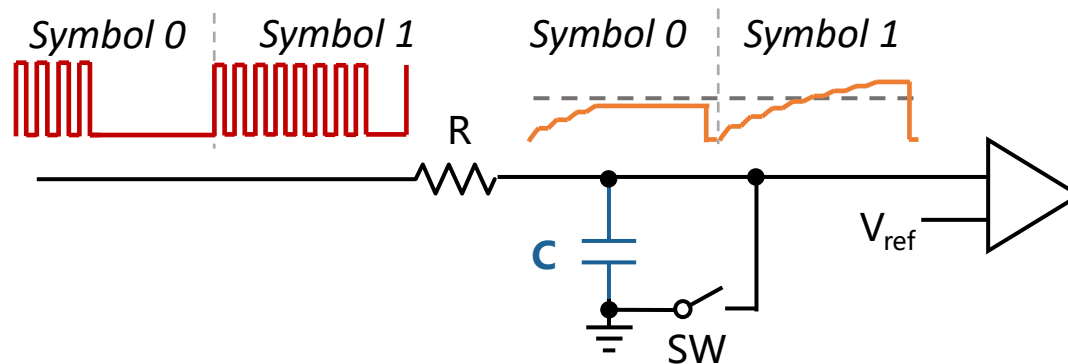
A **symbol** to be decoded



charge



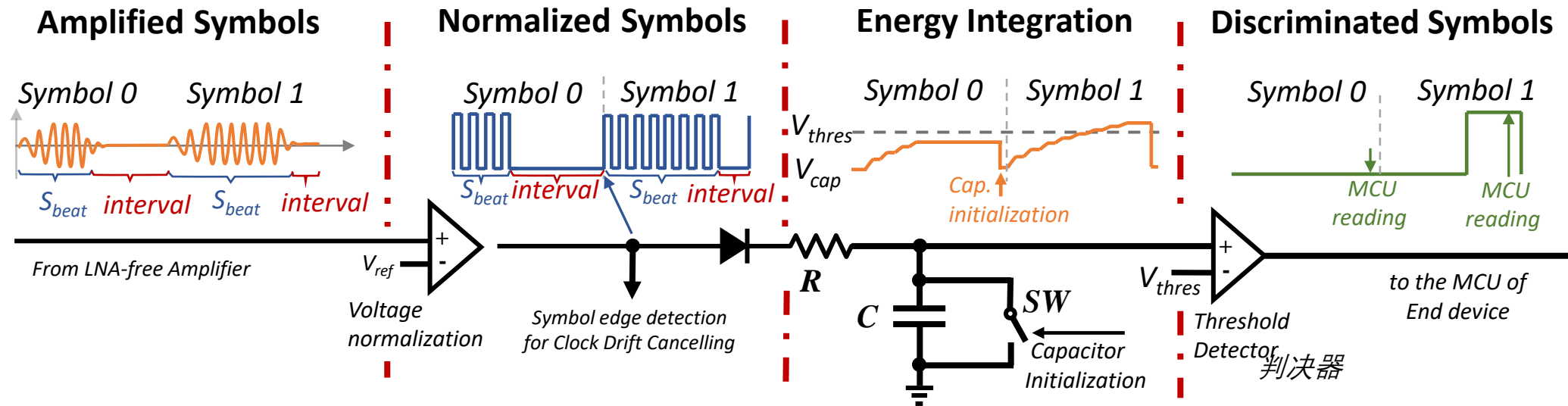
The **signal duration** in the **symbol** will determine the peak voltage in **Capacitor C**



if Peak voltage in $C > V_{ref}$:
decoding result = 1

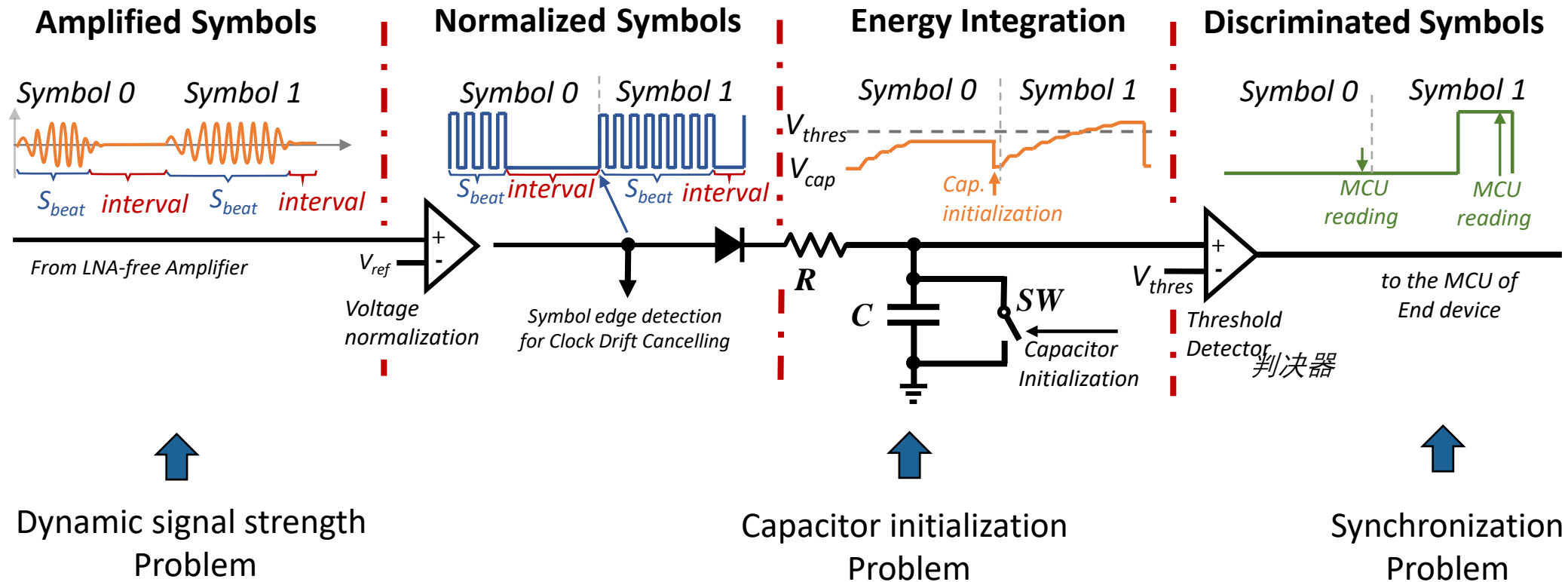
if Peak voltage in $C < V_{ref}$:
decoding result = 0

A more detailed circuit design

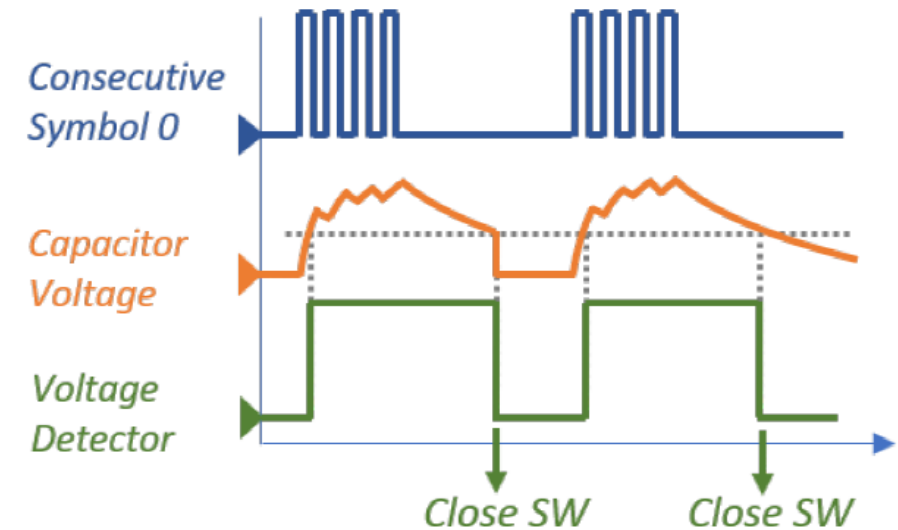
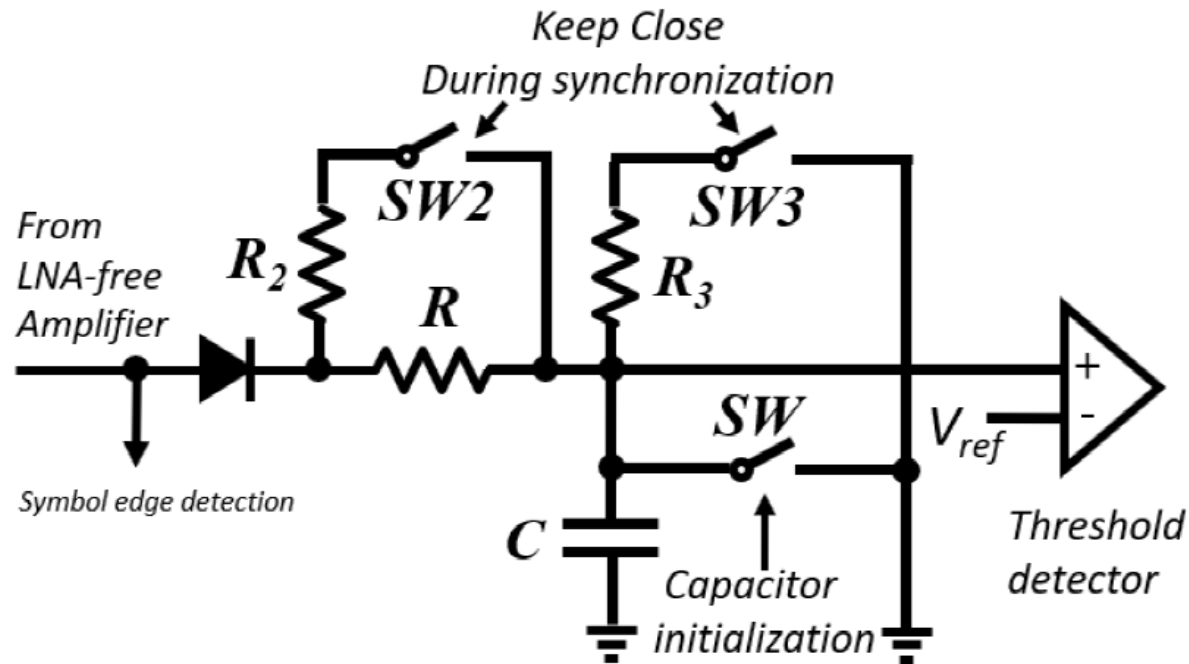


Dynamic signal strength
Problem

A more detailed circuit design

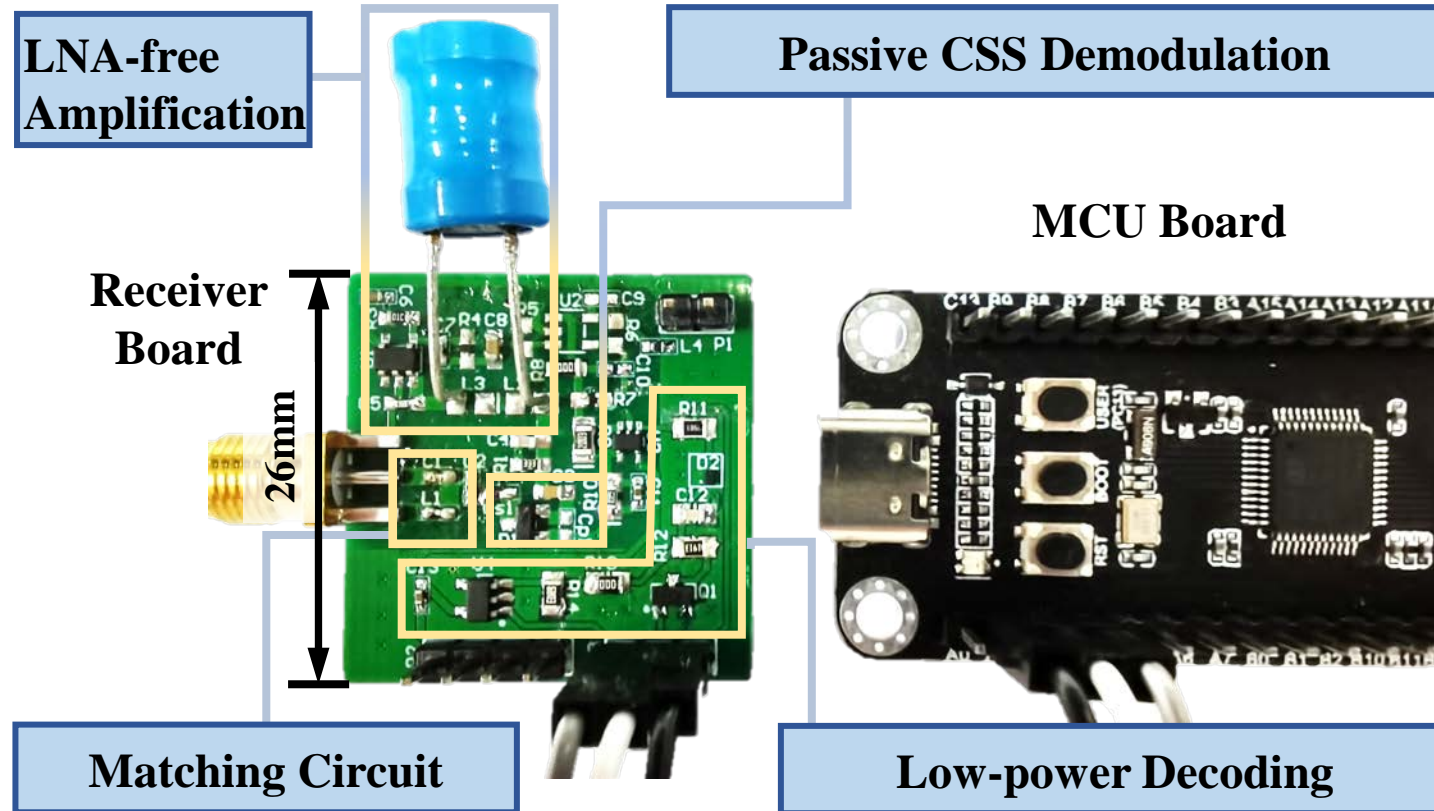


Solution to Synchronization Problem



Change the RC value of the circuit in the synchronization period, making the synchronization symbols (which are actually symbol 0) be able to detected

Implementation



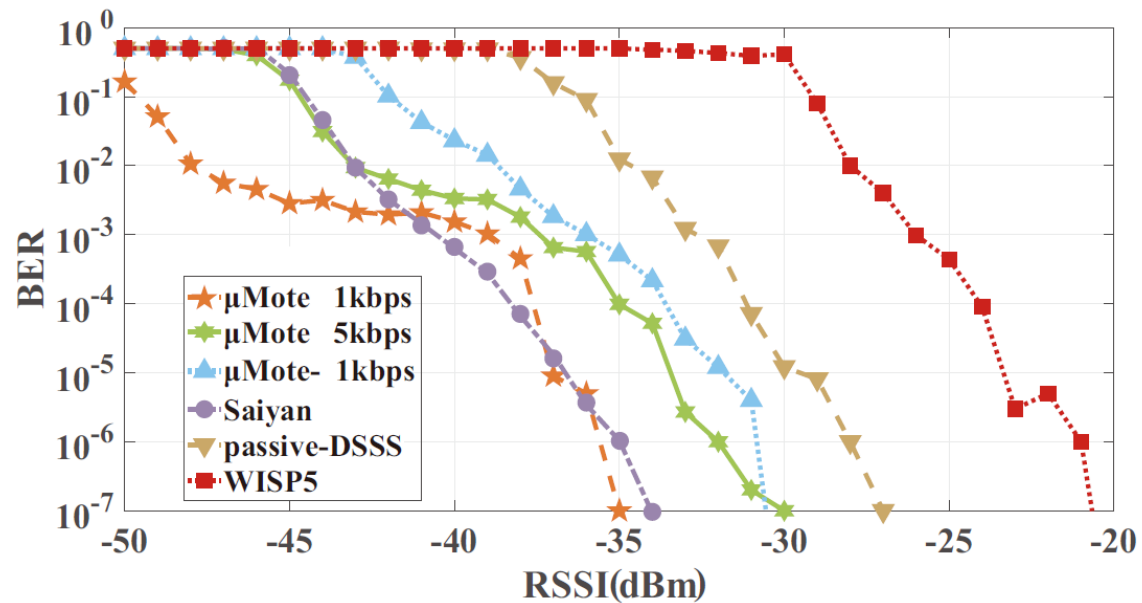
- 26mm × 24mm PCB
- **Power Consumption**
 - “μMote” : 62.07μW
 - “μMote –” : 28.9μW
- Communication rate:
1/2/5kbps

“μMote –” is a μMote prototype which removes the operational amplifier

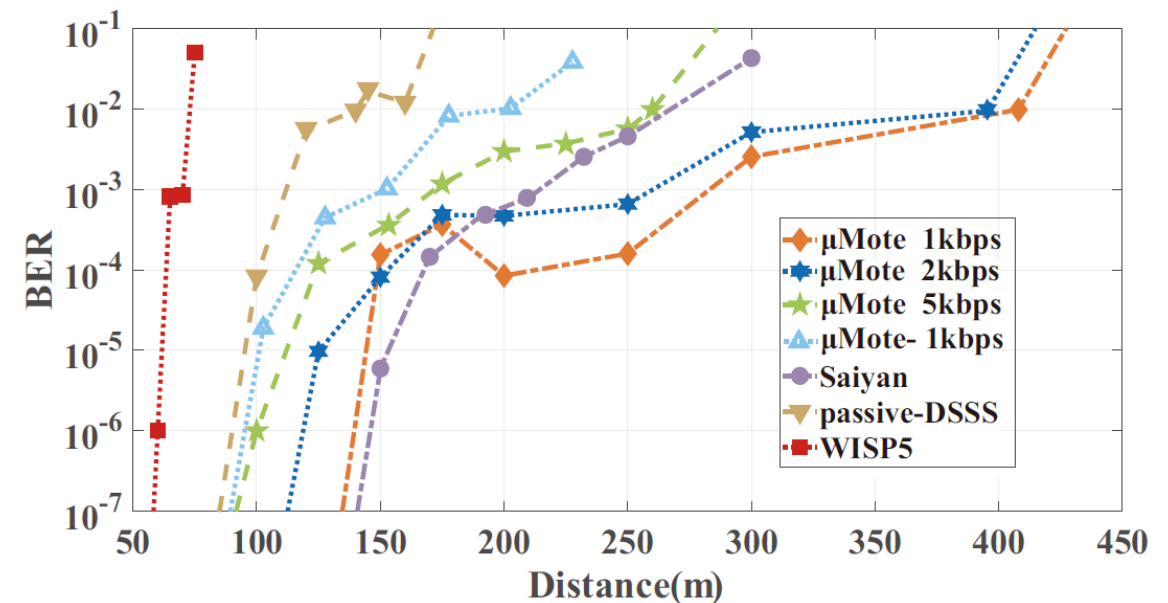
Evaluation

Receiving Sensitivity

(Calibrated by Agilent E4418b RF power meter)



LOS Receiving Range



Benchmarks:

Saiyan¹, passive-DSSS², and WISP5³ (with battery)

¹Saiyan: Design and implementation of a low-power demodulator for LoRa backscatter systems

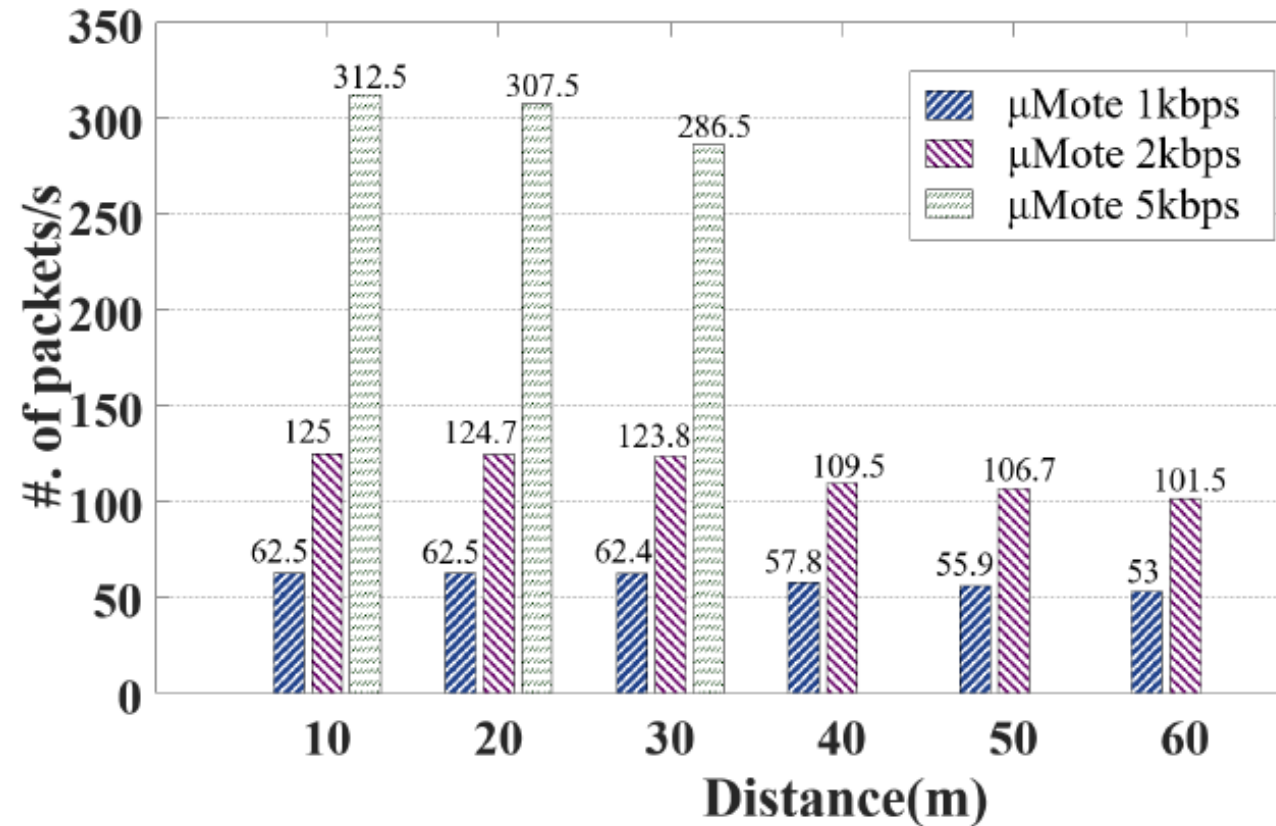
²Passive DSSS: Empowering the downlink communication for backscatter systems

³WISP5: <https://github.com/wisp/wisp5-hw>

Evaluation

NLOS throughput: signal penetrates one wall

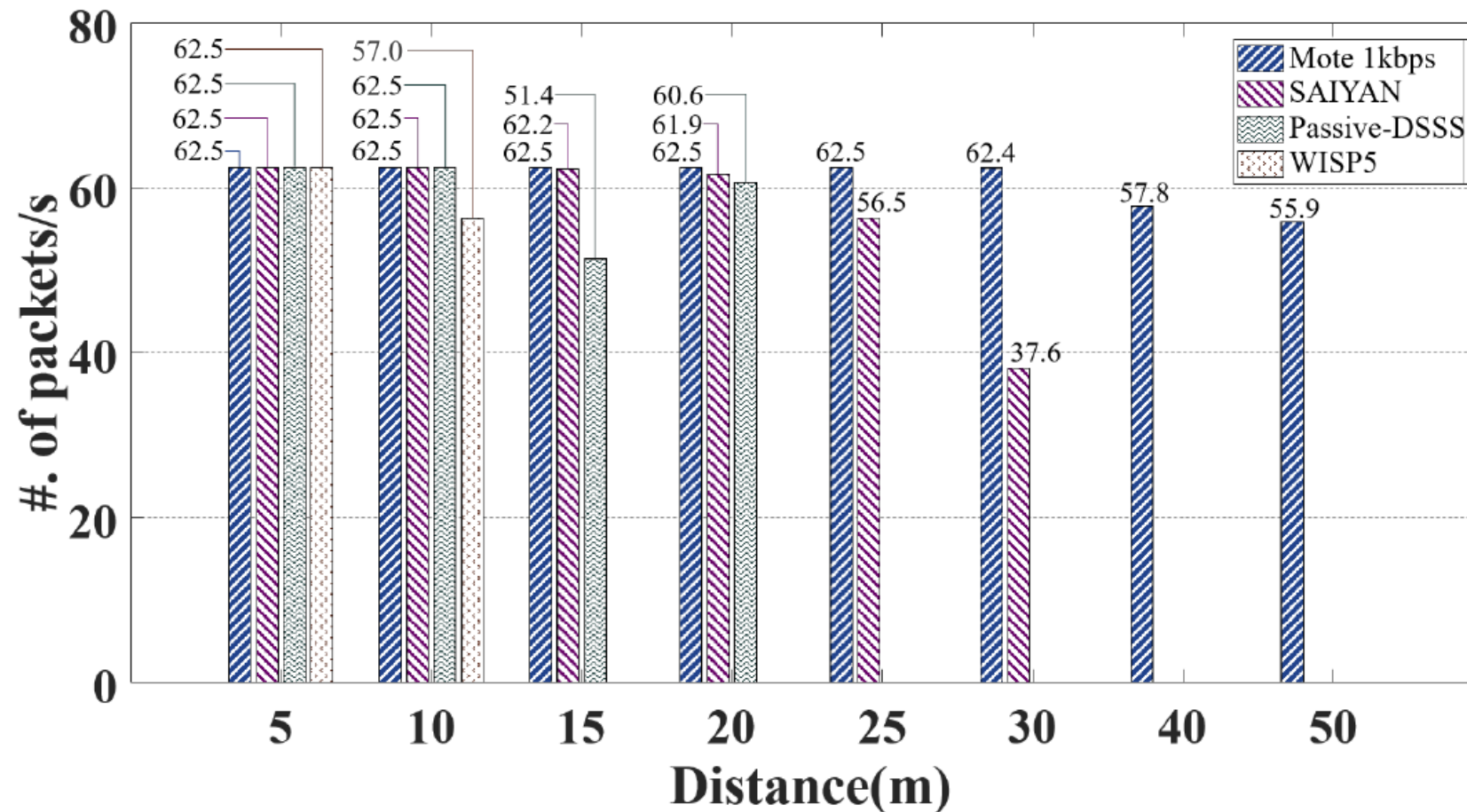
μ Mote @1kbps, 2kbps and 5kbps



Evaluation

NLOS throughput: signal penetrates one wall

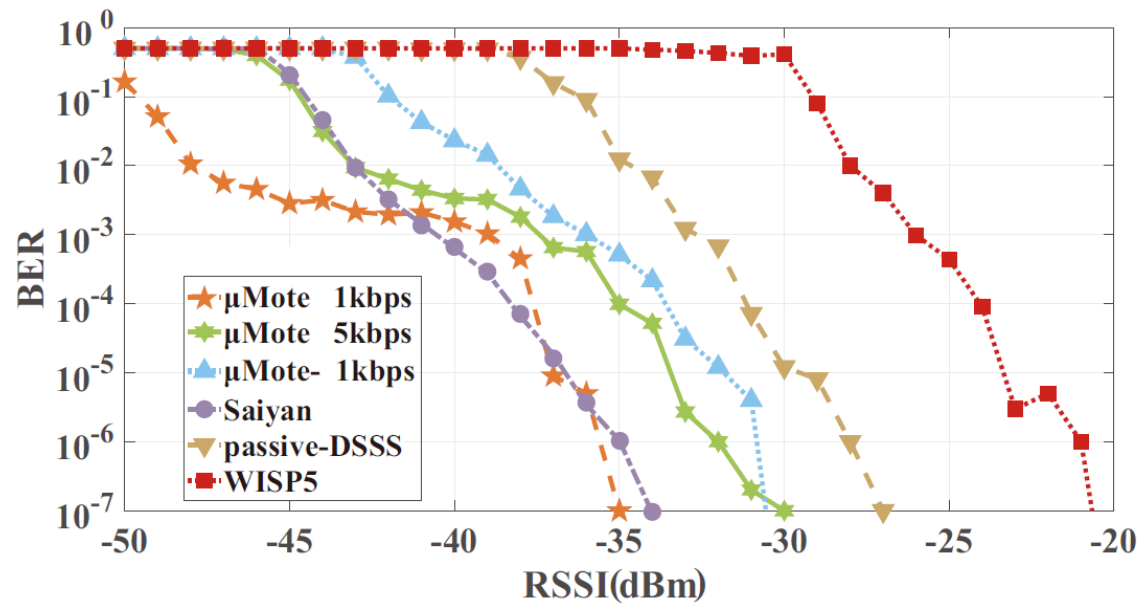
μ Mote and benchmarks @1kbps



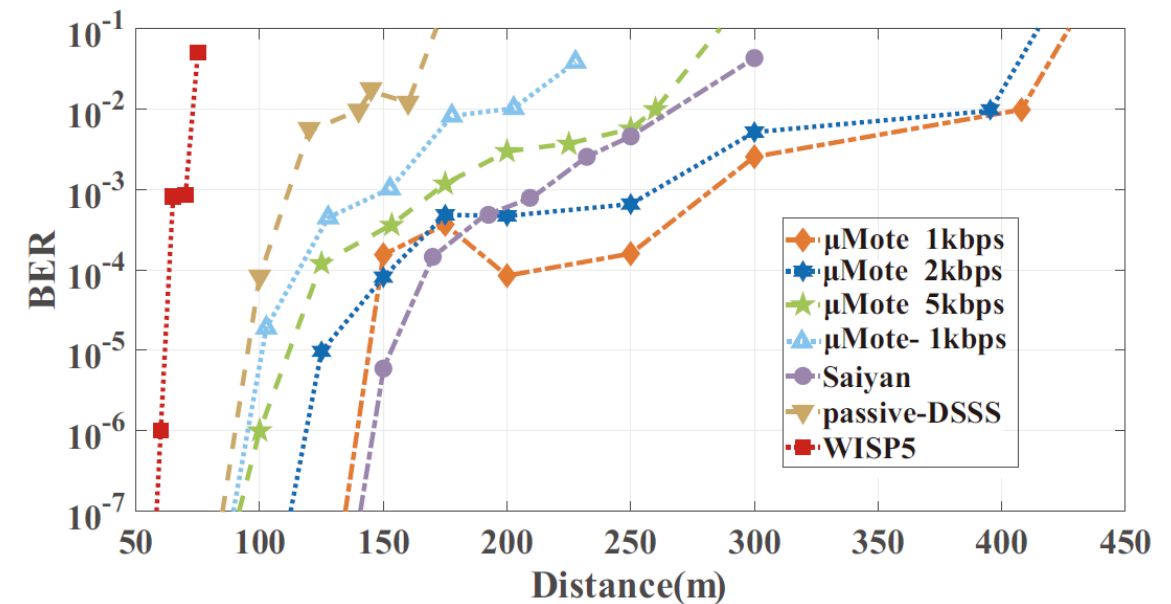
Evaluation

Receiving Sensitivity

(Calibrated by Agilent E4418b RF power meter)



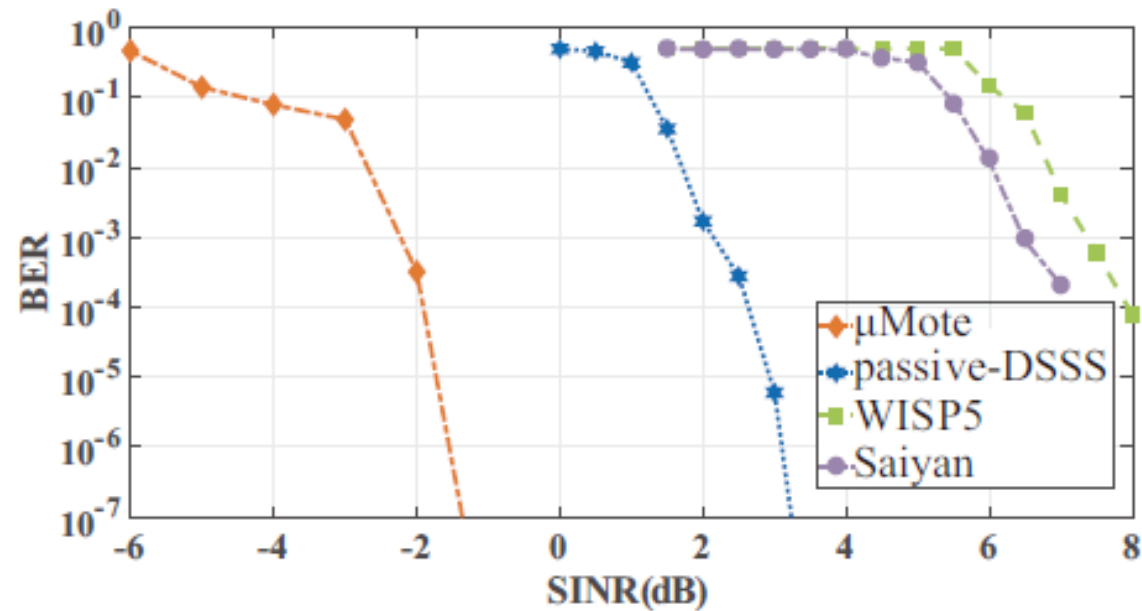
LOS Receiving Range



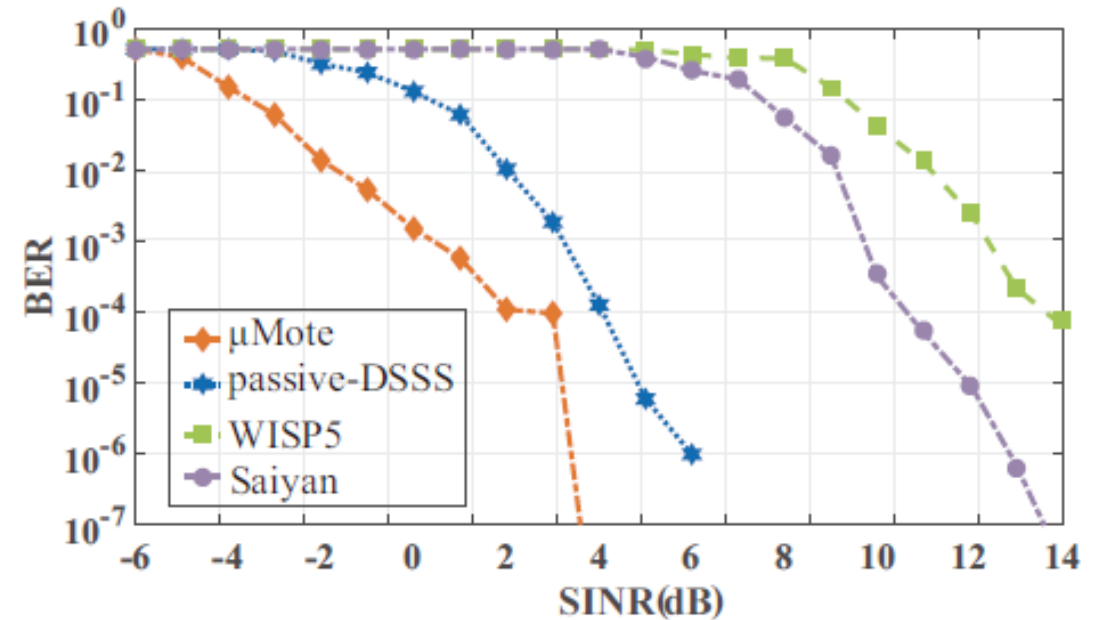
Evaluation

Interference Resistance

RFID interference



LoRa interference



Contributions

We propose the design of μ Mote, a novel μ W-level receiver with hundreds of meters of receiving range which can effectively resist interference and work even under negative SINR.

We address three practical challenges we faced in realizing the receiver, i.e., the passive chirp de-spreading, low-power signal magnification, and low-power decoding.

We prototype μ Mote with COTS components and conduct extensive experiments to verify the feasibility and demonstrate the performance of our design.

Thank You!

 songyihang@uestc.edu.cn